

# Analysis of Settling Time for Low Power CMOS Multistage Operational Amplifiers

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**Abstract** – Design procedure for multistage CMOS op-amp with features of fast settling and low power consumption is present in this paper. This method is focused on optimum compensation by means of proper placement of poles and zero. Single-stage cascode amplifier is no longer suitable in low-voltage designs. So that Multi-stage amplifiers are required with advance in technologies. To reduce the settling time and find the high gain in multi-stage Op-Amp, main aim is minimum mos used in this technology. Nested Miller compensation nulling-resistor technique is used. Simulations on a circuit implemented in a 0.35- $\mu\text{m}$  technology closely to the results expected. Three stage op-amp circuits are simulated by Tanner tool. The results obtained by the circuit simulation are 163 nsec (Settling Time), 90dB (Gain), 9.3 V/ $\mu\text{s}$  (Slew Rate) and 11 MHz (Unity Gain Frequency).

**Keywords** – CMOS, Multistage Amplifiers, Nested-Miller Compensation, Operational Transconductance Amplifiers (OTAs).

## I. INTRODUCTION

The settling time of an op-amp is a very important parameter, particularly for switched-capacitor and data converter circuits. The settling time is defined to be the time takes for op-amp to reach a specified percentage of its final value, when a step input is applied. The total settling time can be broken up into two distinct regions; a slewing period ( $T_{SL}$ ) and a settling period ( $T_{ST}$ ). During the slewing period the output voltage changes from its original value to a voltage close to its final value and the op-amp operates in a rate limited fashion. And during the settling period the output voltage settles to its final value in a small-signal linear fashion. Low voltage, low power cmos multistage operational amplifier is to be designed with compensation technique on the consideration of settling time behavior. Multistage trans-conductance operational amplifiers (OTAs) are increasingly utilized by analog designers since they provide both high gain and voltage swing under reduced voltage supplies. In this paper, we shall describe a design procedure in which the aspect ratio of each transistor is related to the amplifier's main performance parameters. Aspect ratio of the each transistor is optimized according to their best performance. The approach is applied to a three-stage amplifier with low offset, high CMRR performance, high slew rate, high gain(>90 dB), high input impedance and minimum settling time.

The design considerations for fast-settling operational amplifiers (op-amps) are significantly different between sampled-data switched-capacitor (SC) and conventional continuous-time applications. This paper presents a multistage amplifier for low-voltage applications (<3V).

For low voltage supply, multistage amplifier is designed in place of cascode amplifier. The goal of this work is to realize a multi-stage amplifier with a better bandwidth to power efficiency and suitable for driving high capacitive loads such as high accuracy sigma-delta modulators, pipeline A/D converters, linear regulators, and switch capacitors etc. System will be unstable in multistage amplifier design. So many compensation techniques are used to provide the stability, known as Miller compensation techniques. In this paper nested Miller compensation nulling-resistor technique is used to compensate the positive zero introduce in the circuit.

## II. DESIGN PROCEDURE

The design procedure of three-stage amplifier is well known described in this paper. We shall describe a simple and well-defined design procedure in which the aspect ratio of each transistor is related to the amplifier's main performance parameters. Three-stage amplifier is designed on the basis of 0.35 $\mu\text{m}$  technology. The approach is based on the following main parameters: phase margin ( $m$ ), gain-bandwidth product ( $f_{GBW}$ ), load capacitance ( $C_L$ ), slew rate (SR), input common mode range (CMR), input impedance and settling time. Important parameters such as DC gain, CMRR, will not be used during the design steps since they depend on the output resistance of MOS transistors that is not easily modeled.

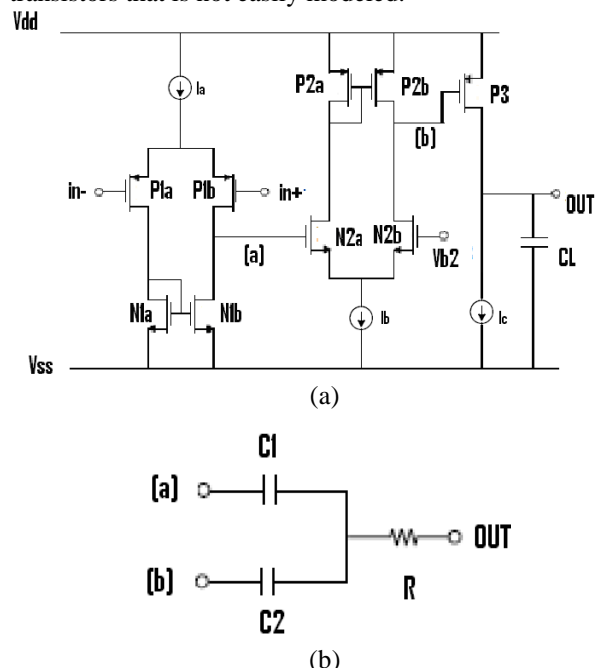


Fig.1. Schematic of the three-stage CMOS OTA (a) and Compensation network (b).

The approach is applied to a three stage op-amp with low voltage, low power for good settling time performance. The circuit is made up of an input differential stage (transistors P1, N1 and current source), a second differential stage (P2, N2 and current source) and a final common-source stage (P3 and current source). Current source is further implemented by cmos current mirror. Parasitic capacitances will be neglected during design because they are much lower than compensation and load capacitances. Moreover, a pencil-and-paper design suggests how to perform modifications if targets have not been reached. By using the current equation W/L can be calculated for all the MOS transistors.

### III. SIMULATION RESULTS

The design is simulated by Tanner tool, using the model parameters of a 0.35- $\mu\text{m}$  double-metal double-poly process with transistors' threshold voltages of about 600mV. The following design parameters were assumed:  $V_{DD}-V_{SS}=2\text{V}$ ,  $C_L=10\text{pF}$ ,  $SR=10\text{V}/\mu\text{s}$ ,  $f_{GBW}=10\text{MHz}$ ,  $m_0=60^\circ$ . We found current through the branches  $I_a=70\mu\text{A}$ ,  $I_b=63\mu\text{A}$  and  $I_c=233\mu\text{A}$ . Moreover, we obtained  $(W/L)_{P1a}=(W/L)_{P1b}=94/1$ ,  $(W/L)_{N1a}=(W/L)_{N1b}=1.6/1$ ,  $(W/L)_{N2a}=(W/L)_{N2b}=45/1$ ,  $(W/L)_{P2a}=(W/L)_{P2b}=90/2$ ,  $(W/L)_{P3}=246/1$ . Regarding the compensation network, we get  $C1=7\text{pF}$ ,  $C2=5\text{pF}$  and  $R=550$ . We simulated a first version of the circuit in Fig. 1 by using the above dimensions and  $I_B=35\mu\text{A}$ . With these settings the following trans conductance values were found:  $gm_{P1}=294\mu\text{A}/\text{V}$ ,  $gm_{N1}=58\mu\text{A}/\text{V}$ ,  $gm_{N2}=371\mu\text{A}/\text{V}$  and  $gm_{P3}=1335\mu\text{A}/\text{V}$ , which, as expected, are different from those estimated by the procedure, causing the circuit's performance to deviate from specifications. Indeed, although the circuit exhibits a very low systematic input offset voltage equal

$$I_D = (1/2) \mu C_{OX} (W/L) (V_{GS} - V_T)^2 \dots\dots\dots (1)$$

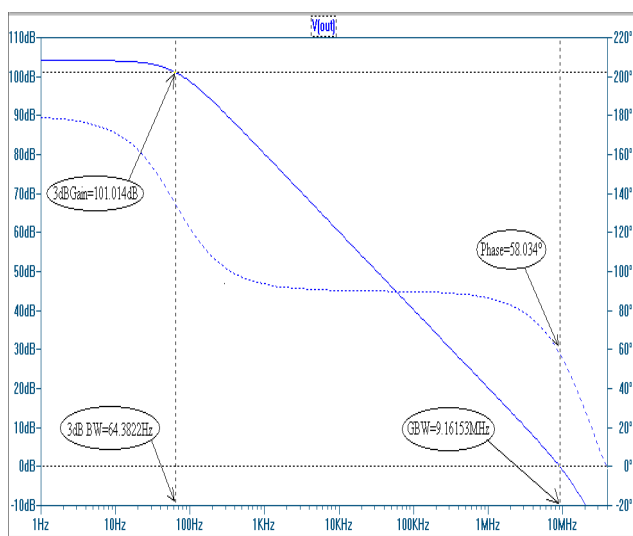


Fig.2. Gain and Phase plot of Op-amp

To design of three stage amplifier, the basic requirement is input resistance to be high and output resistance is less. Hence input differential stage is designed by pmos pair,

because of resistance of pmos is very high as compare to nmos. In input differential stage nmos pair is used as current mirror. If both inputs are to be grounded,  $I_a/2$   $10\mu\text{V}$ , the DC gain and gain-bandwidth product were 87dB and 6.4MHz, respectively with a phase margin of  $60^\circ$ . Positive and negative SR was  $8.9\text{V}/\mu\text{s}$  and  $-9.5\text{V}/\mu\text{s}$ . In particular, the frequency response is illustrated in Fig. 2.

### IV. CONCLUSION

Three-stage Operational Trans-conductance Amplifier has been designed in low-voltage environment. The main research efforts have been aimed on the frequency compensation task, which is perhaps the most difficult one, but that is only one of the several steps required to obtain a complete and efficient amplifier. Well-defined design strategy, closely meeting target specifications should be available. This thesis proposed design procedures for three-stage CMOS OTAs compensated with the simplest possible Nested Miller Compensation Nulling-Resistor technique. Simulated and experimental data from implementations designed in a 0.35- $\mu\text{m}$  technology were found in a reasonable agreement with the results expected. In particular, the op-amp solution exhibits a gain-bandwidth product of 9.16MHz with a SR and settling time of  $13\text{V}/\mu\text{sec}$  and 126.97ns, respectively, while dissipating  $966\mu\text{W}$

Table 1: Simulated Main Performances (Final Design)

Parameters of Op-amp	Values
Input Voltage Range	$\pm 5.8\mu\text{V}$
Power Consumption	$966\mu\text{W}$
Gain	104dB
GBW	9.16MHz
Phase Margin	$120^\circ$
Settling Time	126.97nsec
Slew Rate	$13\text{V}/\mu\text{sec}$
Capacitive Load	10pF

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